

# Sturm-Liouville, PT-symmetric operators and differential algebraic equations

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SOMPATY Lecture

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# Overview

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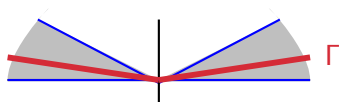
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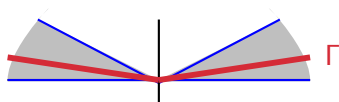
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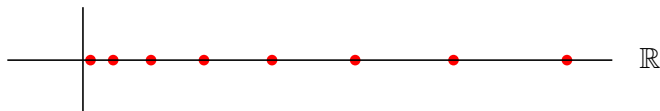


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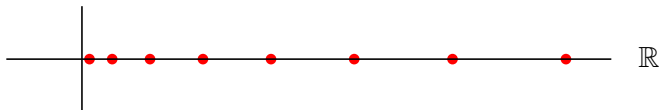


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$f \in L^2$ :

$$f(x) = \frac{2}{\pi} \sum_{n \in \mathbb{N}} b_n \sin nx = \frac{2}{\pi} \sum_{n \in \mathbb{N}} (f, \sin n \cdot)_{L^2} \sin nx \quad (\text{Fourier series})$$

where  $b_n = \int_0^\pi f(t) \sin nt \, dt$ .

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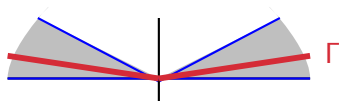
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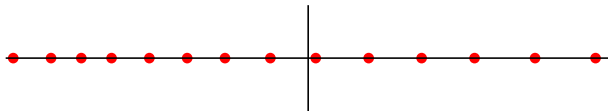
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Then the difference of the resolvents is 1-dim and we obtain

$$\sigma(T) = \sigma_p(T) \subset \mathbb{R}.$$



# Indefinite Sturm Liouville

## Interested in

- the location of point and essential spectrum
- eigenvalue asymptotics
- non-real spectrum
- accumulation of non-real eigenvalues or of eigenvalues in gaps of the essential spectrum



# Indefinite Sturm-Liouville, more difficult

Theorem (Math. Ann.'13)

Let  $r := \text{sgn}$ ,  $p = 1$ , and  $q \in L^\infty(\mathbb{R})$ ,  $\text{ess\,inf } q < 0$ ,

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Then the non-real spectrum of  $T$  is contained in

$$\left\{ \lambda \in \mathbb{C} : \text{dist}(\lambda, (-d, d)) \leq 5\|q\|_\infty, |\text{Im}\lambda| \leq 2\|q\|_\infty \right\},$$

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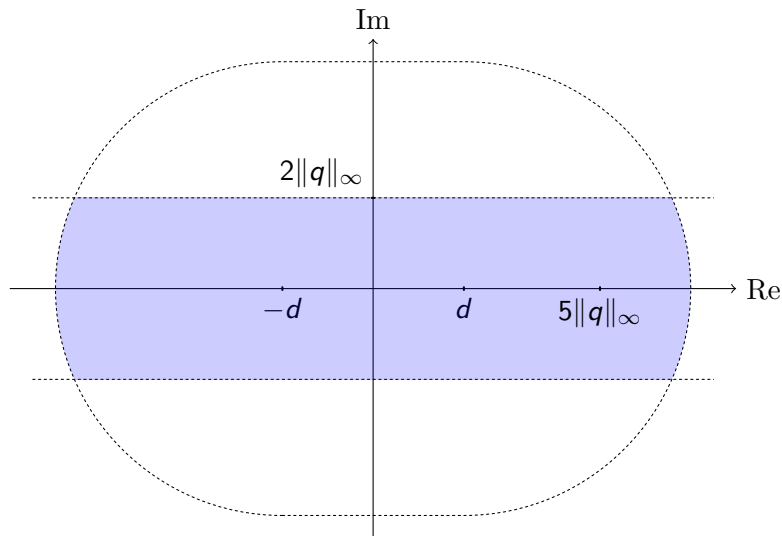
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## Indefinite Sturm-Liouville, more difficult II



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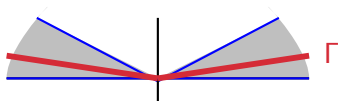
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## 2. Order systems II

Theorem (B. Jacob, C. Tretter, CT, H. Vogt, Math. Meth. Appl. Sci.'18)

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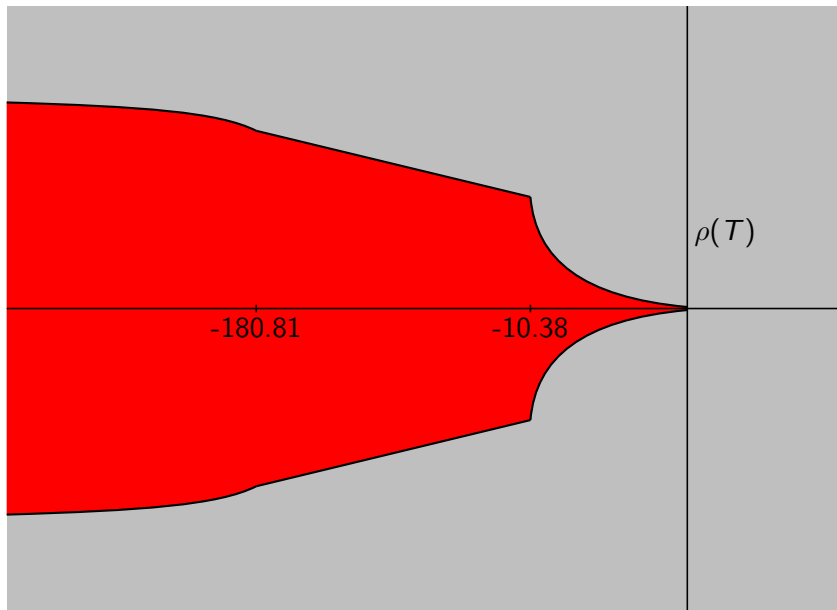
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Proof: With the quadratic numeric range.

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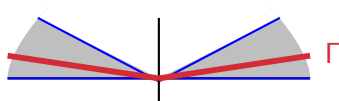
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## Real Spectra in Non-Hermitian Hamiltonians Having $\mathcal{PT}$ Symmetry

Carl M. Bender<sup>1</sup> and Stefan Boettcher<sup>2,3</sup>

<sup>1</sup>*Department of Physics, Washington University, St. Louis, Missouri 63130*

<sup>2</sup>*Center for Nonlinear Studies, Los Alamos National Laboratory, Los Alamos, New Mexico 87545*

<sup>3</sup>*CTSPS, Clark Atlanta University, Atlanta, Georgia 30314*

(Received 1 December 1997; revised manuscript received 9 April 1998)

The condition of self-adjointness ensures that the eigenvalues of a Hamiltonian are real and bounded below. Replacing this condition by the weaker condition of  $\mathcal{PT}$  symmetry, one obtains new infinite classes of complex Hamiltonians whose spectra are also real and positive. These  $\mathcal{PT}$  symmetric theories may be viewed as analytic continuations of conventional theories from real to complex phase space. This paper describes the unusual classical and quantum properties of these theories. [S0031-9007(98)06371-6]

The Hamiltonian studied by Bessis is just one example of a huge and remarkable class of non-Hermitian Hamiltonians whose energy levels are real and positive. The purpose of this Letter is to understand the fundamental properties of such a theory by examining the class of quantum-mechanical Hamiltonians

$$H = p^2 + m^2 x^2 - (ix)^N \quad (N \text{ real}). \quad (1)$$

As a function of  $N$  and mass  $m^2$  we find various phases with transition points at which entirely real spectra begin to develop complex eigenvalues.



### ③ Non Hermitian quantum mechanics

$$\ell(y) := -y''(z) - (iz)^{N+2}y(z), \quad N \in \mathbb{N}$$

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### Problems:

- ① Potential is complex valued.
- ②  $\Gamma$ .
- ③ For eigenvalues we need operators. Which? Domains?
- ④ What is  $\mathcal{PT}$  symmetric ? How it will help us?

## $\mathcal{PT}$ symmetric expressions

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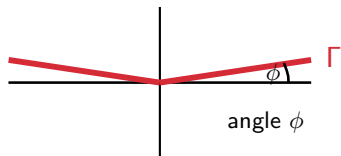
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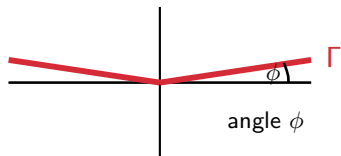
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We have

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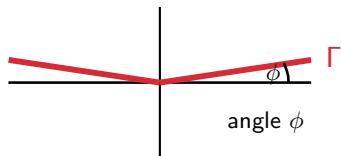
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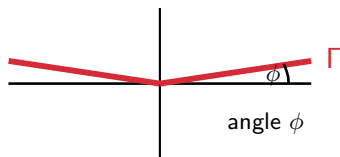
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hence (formally)

$$\ell\mathcal{PT} = \mathcal{PT}\ell$$

and  $\ell$  is called  $\mathcal{PT}$  symmetric.

# Overview

- 1 Indefinite Sturm-Liouville:

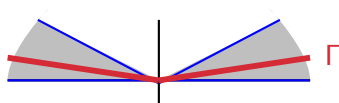
$$T_y := \frac{1}{r} (-(py')' + qy).$$

- 2. Order systems:

$$\ddot{u} = -30u_{\xi\xi\xi\xi} - 3\dot{u}_{\xi\xi\xi\xi} - u_{\xi t}.$$

- 3 Non Hermitian quantum mechanic:

$$T_y(z) := -y''(z) + z^2(iz)^\epsilon y(z), \quad \epsilon > 0, \quad z \in \Gamma.$$



- 4 Differential algebraic equations (DAE):

$$E\dot{x} = Ax, \quad \text{in chip re-design: Perturb } E.$$

# Simple Example

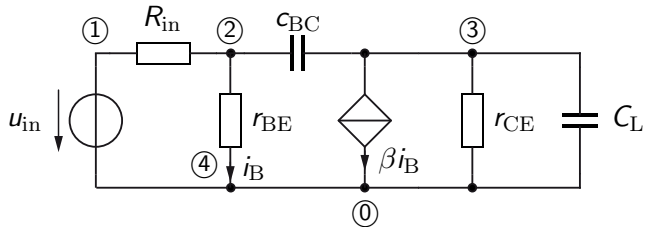


Figure: Amplifier

Laplace transform

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Resistor

$$i = \frac{V}{R}$$

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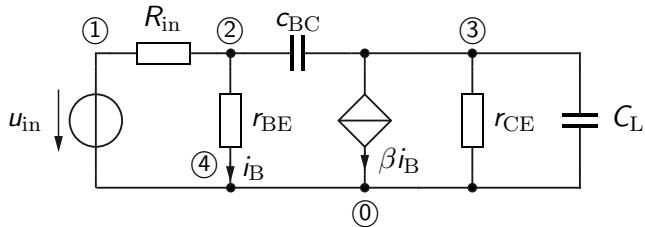


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Resistor	$i = \frac{V}{R}$	$i = \frac{V}{R}$
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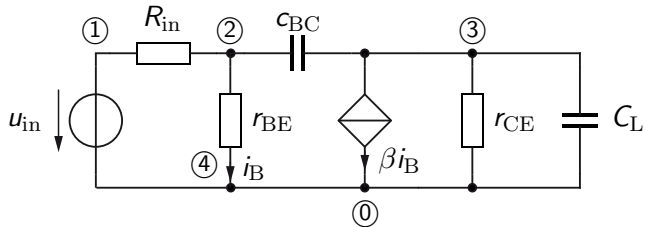


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Laplace transform

Resistor  $i = \frac{V}{R}$   $i = \frac{V}{R}$

Capacitance  $i = C \cdot \dot{V}$   $i = C \cdot sV$

Current source  
(current controlled)  $i = \beta i_B$   $i = \beta i_B$



# Simple Example

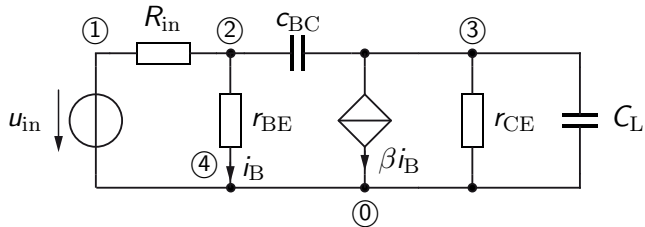


Figure: Amplifier

$$\begin{pmatrix} \frac{1}{R_{in}} & -\frac{1}{R_{in}} & 0 & 0 & 1 & 0 \\ -\frac{1}{R_{in}} & \frac{1}{R_{in}} + \frac{1}{r_{BE}} + C_{BC}S & -C_{BC}S & -\frac{1}{r_{BE}} & 0 & 0 \\ 0 & -C_{BC}S & \frac{1}{r_{CE}} + C_{BC}S + C_L S & 0 & 0 & \beta \\ 0 & -\frac{1}{r_{BE}} & 0 & \frac{1}{r_{BE}} & 0 & 1 \\ 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \\ V_3 \\ V_4 \\ i_{in} \\ i_B \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \\ 0 \\ u_{in} \\ 0 \end{pmatrix}$$

# DAE

$$\underbrace{\begin{pmatrix} \frac{1}{R_{in}} & -\frac{1}{R_{in}} & 0 & 0 & 1 & 0 \\ -\frac{1}{R_{in}} & \frac{1}{R_{in}} + \frac{1}{r_{BE}} + C_{BC}s & -C_{BC}s & -\frac{1}{r_{BE}} & 0 & 0 \\ 0 & -C_{BC}s & \frac{1}{r_{CE}} + C_{BC}s + C_{L}s & 0 & 0 & \beta \\ 0 & -\frac{1}{r_{BE}} & 0 & \frac{1}{r_{BE}} & 0 & 1 \\ 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \end{pmatrix}}_{E s - A} \begin{pmatrix} V_1 \\ V_2 \\ V_3 \\ V_4 \\ i_{in} \\ i_B \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \\ 0 \\ u_{in} \\ 0 \end{pmatrix}$$

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How to change the modulus of the *transfer function*  $H$  on  $j\mathbb{R}$

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through a new capacitance  $c$ ?

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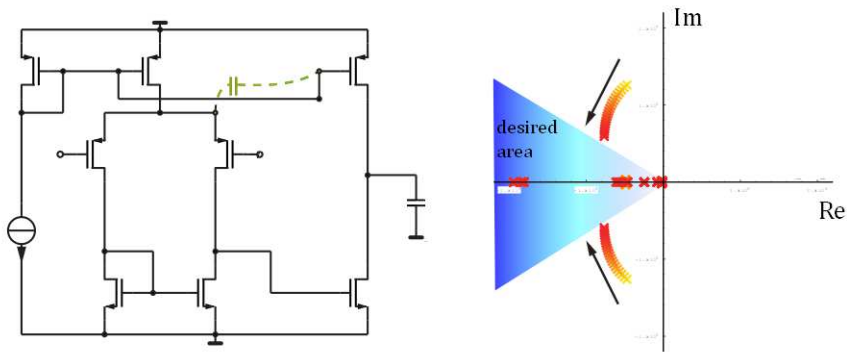
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New capacitance:

$$E \rightarrow E + cpp^T \quad \text{with} \quad p = e_j - e_k.$$

# Network redesign (Sommer, Krauß and others)

- **Method:** Add step-by-step capacitances to the network between the nodes  $i$  and  $j$ , described by the pencil  $sc_{ij}(e_i - e_j)(e_i - e_j)^T$ ,  $c_{ij} > 0$ .
- **New matrix pencil:**  $s(E + c_{ij}(e_i - e_j)(e_i - e_j)^T) - A$



**Thank you.**